WIDEBAND MICROSTRIP-FED PRINTED BOW-TIE ANTENNA FOR PHASEDARRAY SYSTEMS

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ABSTRACT: A micro strip-fed printed bow-tie antenna is presented inorder to achieve wide bandwidth, high gain, and size reduction. A comparisonbetween the bow-tie and the quasi-Yagi (dipole and director)antennas shows that the bow-tie antenna has a wider bandwidth, highergain, lower front-to-back ratio, lower cross-polarization level, and smaller size. Two-element arrays are designed and their characteristics are compared. The bow-tie antenna yields lower coupling for the same distance between elements.

Key words: wideband antennas; printed bow-tie antenna; printed quasi-Yagi antenna

1. INTRODUCTION

Printed microstrip antennas are widely used in phased-array applicationsbecause they exhibit a very low profile, small size, lightweight, low cost, high efficiency and easy methods of fabricationand installation. Among the most widely used printed antennas inphased-array systems are printed dipoles and quasi-Yagi antennasfed by coplanar strip line (CPS), which are usually used to yieldend-fire radiation patterns. In order to feed this antenna, someresearchers suggest microstrip-to-CPS transition that includes a180° phase shifter [1]. Other researchers feed the dipole with twomicrostrip lines where the upper is an extension of the microstripfeed line and lower is connected to the ground plane directly orthrough a tapered microstrip [2, 3]. However, the latter methodssuffer from low radiation efficiency (88% in [2]) and low bandwidth(37% in [2] and 19% in [3]). Moreover, unbalanced radiationpatterns are noticed in [2] and omnidirectional patterns areobtained in [3]. Other researchers use coplanar waveguide (CPW) -to-CPS transitions to feed printed dipole and bow-tie antennas [4]. However, these two antennas are designed for 100_, not 50_, characteristic impedance, in addition to having an omnidirectional pattern. An attractive design that uses the transition in [1] is presented in [5, 6] and exhibits wide bandwidth and good radiation charac-teristics.

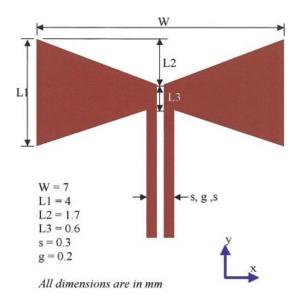


Figure 1 Geometry and dimensions of the printed bow-tie antenna

The antenna consists of a half-wavelength dipole and anapproximately quarter-wavelength rectangular director in order toincrease the gain and improve the front-to-back ratio. In this paper, the printed dipole and the director of [5, 6] are replaced by aprinted bow-tie, which results in an improvement in bandwidth andgain. That is because printed bow-tie antennas are planar-typevariations of the biconical antenna that has wideband characteristics. Moreover, the radiating area of the bow-tie is larger than that of the dipole; therefore, gain

improvement is expected. The simulationand analysis for this new antenna are performed using thecommercial software package an soft HFSS, which is based on the finite-element method. The measurements of the return loss and radiation pattern are also conducted.

2. SINGLE ELEMENT

The proposed antenna element is printed on a Rogers RT/Duroid6010/6010 LM substrate with a dielectric constant of 10.2, athickness of 25 mil, and a conductor loss (tan _) of 0.0023. Themicrostrip-to-CPS transition is almost the same as that in [1]. Thebow-tie geometry and dimensions are shown in Figure 1. Thequasi-Yagi antenna [5, 6] is simulated in order to compare it withthe new bow-tie design on the same material-type substrate andground-plane dimensions. The simulated and measured return losses of the bow-tie antenna, compared to those of the quasi-Yagi, are shown in Figures2 and 3, respectively. According to the HFSS simulation results, the bow-tie shows about 13% improvements in the bandwidth, where it operates from 6.8 to 11.9 GHz with a bandwidth of54.5%, while the quasi-Yagi operates from 7.9 to 12.1 GHz, witha bandwidth of 41.6%. In the measurements, the bow-tie showsabout 19.6% improvements in the bandwidth, where it operates from 6.7 to 12.45 GHz with a bandwidth of 60.1%, while thequasi-Yagi operates from 8.2 to 12.5 GHz, with a bandwidth of41.5%. The copolarized (*E*_) and cross-polarized (*E*_) far-field radiation patterns for the two antennas are computed at 10 GHz. Figure4 shows the radiation patterns of the bow-tie antennas, while Figure 5 shows the radiation pattern of the quasi-Yagi antenna.

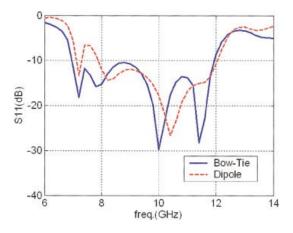


Figure 2 Computed return losses of the bow-tie and the quasi-Yagi Antennas.

The simulation results show that at least 1.3-dB improvement in the gain has been obtained when using the bowtie. The maximumgain for the bow-tie is around 5.7 dB, while it is around 4.4 dB for the quasi-Yagi. The 3-dB beam width in the E-plane (x-y) is almost the same for both antennas: 106° and 108° for the bow-tie and the quasi-Yagi antennas, respectively. However, in the Hplane(y-z), the quasi-Yagi shows much wider beam width: 108° for the bow-tie and 153° for the quasi-Yagi antenna. The H-planepattern becomes more focused for the bow-tie, which results inenhanced gain and reduced beam width. As shown in Figures 4 and 5, the computed front-to-back ratio is improved by 1.5 dB, whereit is around 14.1 dB for the bow-tie and 12.6 dB for the quasi-Yagi. The cross-polarization level in the E-plane is 22.5 dB for thebow-tie, while it is 20 dB for the quasi-Yagi, and for the H-planeit equals to 23 dB for the bow-tie and 24 dB for the quasi-Yagi, considering only the angles defining by the 3-dB beam width.

3. TWO-ELEMENT ARRAY

Two elements of the bow-tie and quasi-Yagi antennas are simulated and fabricated in order to compare the coupling (S21 in dB)between the array elements. The distance between elements is fixed to 15 mm, which is the free-space half-wavelength at 10

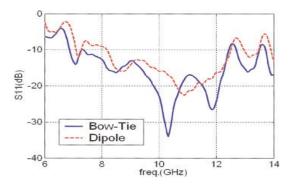


Figure 3 Measured return losses of the bow-tie and the quasi-Yagi antennas.

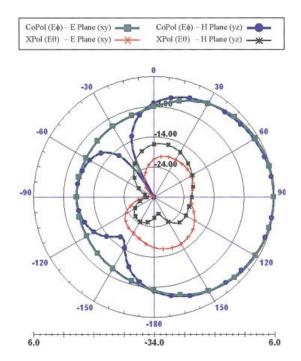


Figure 4 Computed far-field radiation pattern for the bow-tie antenna at 10 GHz.

GHz. Photographs of the two-element arrays are shown in Figure 6. Figure 7 shows a comparison of the measured coupling betweenthe bow-tie and quasi-Yagi elements. The coupling is less betweenthe bow-tie elements, as shown in Figure 7, where the couplingimproves by an average value of around 4 dB It is worth mentioning

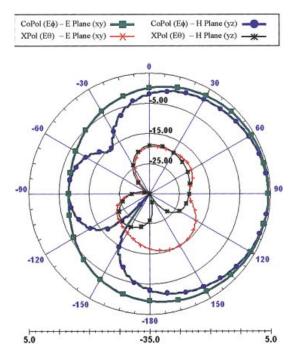


Figure 5 Computed far-field radiation pattern for the quasi-Yagi antennaat 10 GHz.

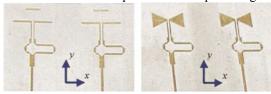


Figure 6 Photograph of a two-element array of the bow-tie and quasi- Yagi antennas.

That this improved coupling is also associated with antenna-size reduction, as the bow-tie edge-to-edge dimension is 7 mmwhile that of the quasi-Yagi is 8.7 mm, which gives a 24% reduction. The co- and cross-polarized far-field radiation patterns fortwo-element arrays of the bow-tie and quasi-Yagi antennas are computed at 10 GHz. Figure 8 shows the radiation patterns of thetwo-element array of the bow-tie antenna, while Figure 9 shows the radiation pattern of the two-element array of the quasi-Yagiantenna. According to these results, approximately 2-dB improvement in the gain has been obtained with the bow-tie array. Themaximum gain for the bow-tie array is around 9.3 dB, while it isaround 7.3 dB for the quasi-Yagi array. The 3-dB beam width ofthe co-polarized pattern in the E-plane is 46° and 48° for thebow-tie and the quasi-Yagi, respectively. The beam width in theH-plane for the quasi-Yagi is 120°, while that for the bow-tie it is90°; this are different from that of the one-element configuration due to the coupling between the elements. The front-to-back ratiois also found to be improved, as it is 20.7 dB for the bow-tieantenna array and 11.7 dB for the quasi-Yagi antenna array. Thecross-polarization level is also enhanced using bow-tie elements. In the E-plane, the cross polarization level is -29 dB for thebow-tie while it is -26 dB for the quasi-Yagi, and for the H-planeit equals to -26 dB for the bow-tie and -24 dB for the quasi-Yagi.

4. CONCLUSION

In this paper, a printed bow-tie antenna has been designed toreplace the dipole and the director in the printed quasi-Yagiantenna configuration. This new bow-tie design provides widerbandwidth, smaller size, higher gain, and smaller cross polarizationthan the quasi-Yagi, and shows an improvement in the front-to-backratios for one-and two-element arrays. The design oflarger arrays based on this type of antenna is therefore moreappropriate for phased-array systems.

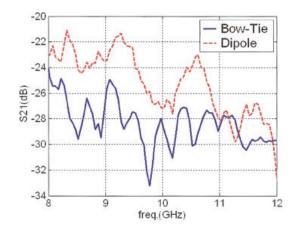


Figure 7 Comparison of the measured coupling for two-element arraysof the bow-tie and quasi-Yagi antennas.

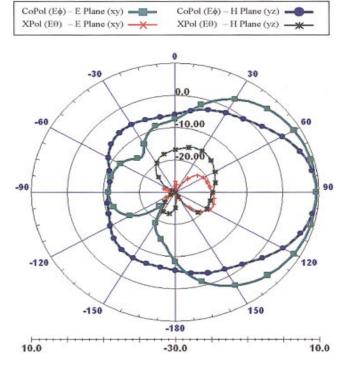


Figure 8 Computed far-field radiation pattern for the two-element arrayof the bow-tie antenna at 10 GHz.

REFERENCES

- 1. N. Kaneda, Y. Qian, and T. Itoh, A broadband microstrip-to-waveguidetransition using quasi-Yagi antenna, IEEE Trans Microwave TheoryTech 47 (1999), 2562–2567.
- 2. G. Zheng, A. Kishk, A. Yakovlev, and A. Glisson, Simplified feedingfor a modified printed Yagi antenna, IEEE Antennas Propagat Soc IntSymp Dig 3 (2003), 934–937.
- 3. G.Y. Chen and J.S. Sun, A printed dipole antenna with microstriptapered balun, Microwave Opt Technol Lett 40 (2004), 344–346.
- 4. C.W. Chiu, Coplanar-waveguide-fed uniplanar antenna using a broadbandbalun, Microwave Opt Technol Lett 40 (2004), 70–73.
- 5. W. Deal, N. Kaneda, J. Sor, Y. Qian, and T. Itoh, A new quasi-Yagiantenna for planar active antenna arrays, IEEE Trans Microwave TheoryTech 48 (2000), 910–918.
- 6. N. Kaneda, W. Deal, Y. Qian, R. Waterhouse, and T. Itoh, A broadbandplanar quasi-Yagi antenna, IEEE Trans Antennas Propagat 50 (2002),1158–1160.